

Folded cascode

freq resp, stability

lossy -

Project

Switched capacitors

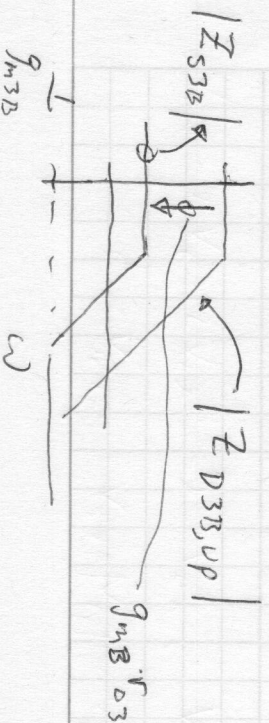
Why is there not a pole at $D2B/SSB$?

Capacitance is C_{g53B}

low freq resistance is $(r_{o2B} || \frac{g_{m4} r_{o4}}{g_{m3} r_{o3}}) r_{o5} \approx r_{o5}$

Why not a pole at $\frac{1}{r_{o5} C_{g53B}}$?

A: $Z_{D2B,up} = (g_{m4} r_{o4}) r_{o5} || \frac{1}{\omega C_L}$



Folded cascode

freq resp.

if you don't like the

phase margin, add

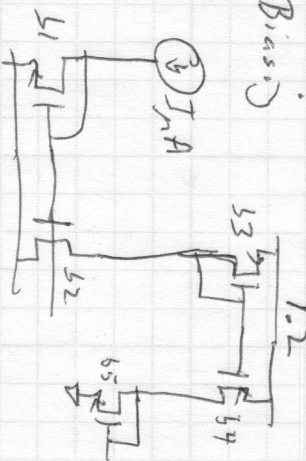
capacitance to C_L

contrast w/ 2 pole:

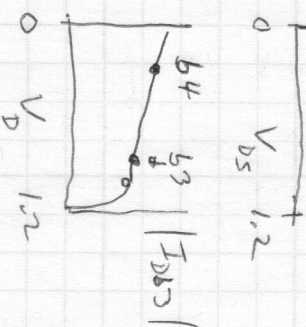
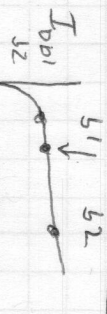
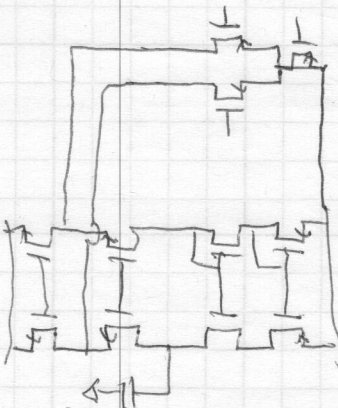
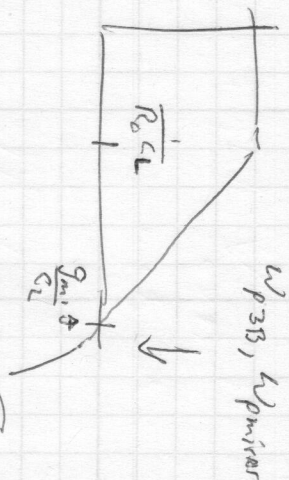
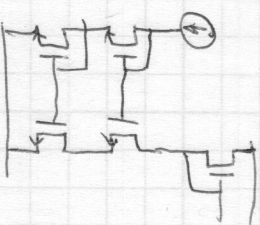
$\omega_{uc} = \frac{g_{m1}}{C_C}$ } less stable

$\omega_{pzc} = \frac{g_{m4}}{C_1 + C_2}$ } w/ more load cap.

Biasing



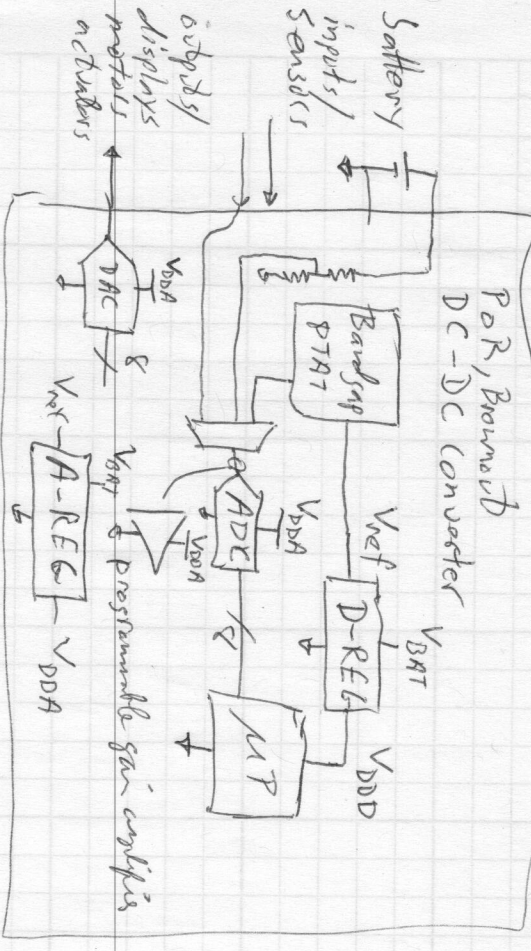
much better:



14/2/2014

1750 W11L3

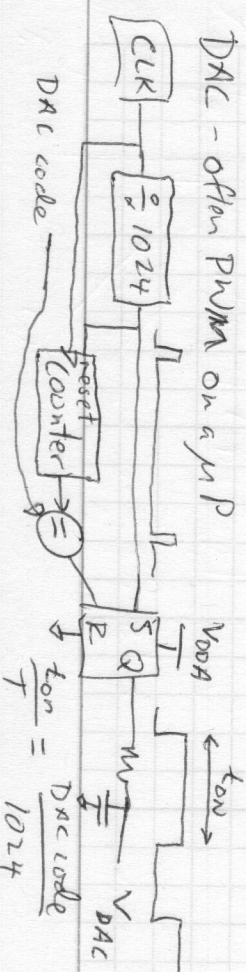
Analogy for
 Embedded MP or IOT SoC, or why do we
 care about op-amps?



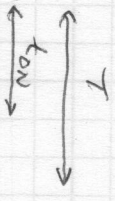
Lots of real-world examples, e.g. HW9
 ADI regulators
 TI ADCs

Dash Networks chips

SCMM-2
 EE194/29DC



DAC - often PWM or a MP



$$t_{on} = \frac{DAC\ code}{1024} \cdot T$$

At least 4 op-amps

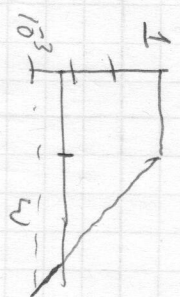
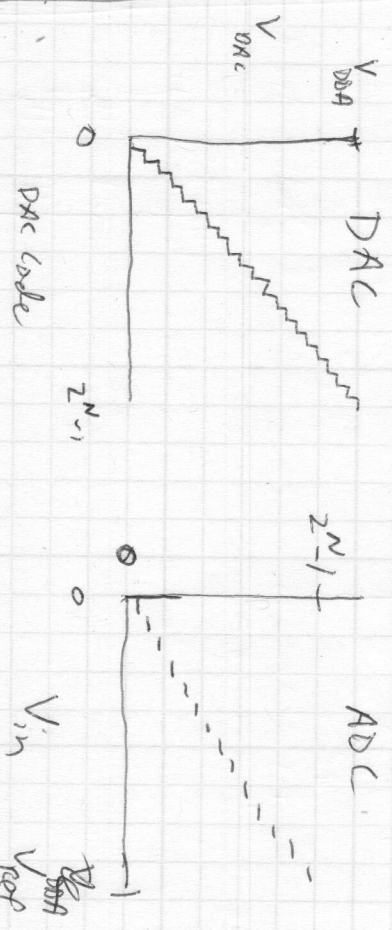
- Bandgap
- AREL
- D-REG
- P-GA

op-amp or comparator

ADC

$$\overline{V}_{DAC} = V_{DDA} \frac{t_{on}}{T}$$

Real $T \ll RC$

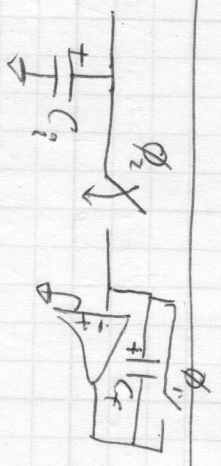


PGA Razavi

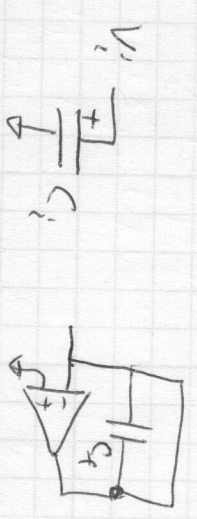
switched capacitors - popularized by Berkeley Gray et al.
 goal: use capacitors instead of resistors
 to capture loads of CMOS

- MOS transistors make good switches
- MOS op-amps also capture loads

e.g. $V_i \frac{\phi_1}{X}$

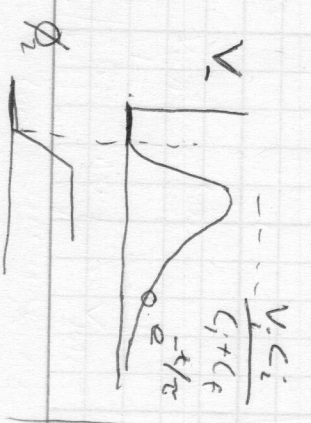
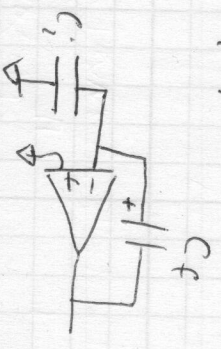


during ϕ_1 :

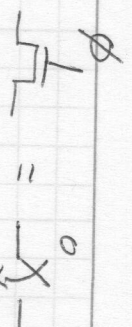


$$Q_{i+} = V_i C_i$$

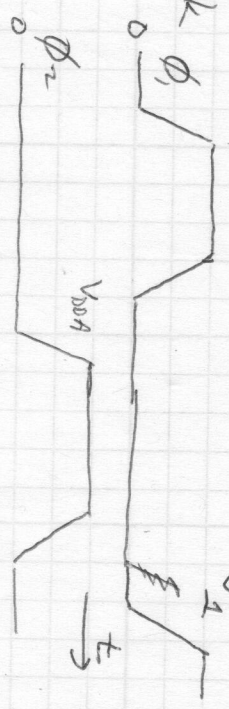
during ϕ_2 :



transistor is a switch



ϕ_1 : clock



often with 2 or more non-overlapping phases.

→ Lets you change the wiring in your circuit

"during ϕ_1 , it looks like"

"during ϕ_2 it looks like"

What's also things settle (note)

must have $V_+ = V_- = 0 \Rightarrow Q_{i+} = 0$

$$\Rightarrow Q_{F+} = V_i C_i \quad (\text{from } \phi_1)$$

$$\Rightarrow V_{CF} = \frac{Q_{F+}}{C_f} = V_i \frac{C_i}{C_f}$$

$$V_{out} = V_- - V_{CF} = 0 - V_i \frac{C_i}{C_f}$$

switched capacitor amplifier n/ gain = $-\frac{C_i}{C_f}$
 can choose C_i with switches \Rightarrow programmable C_i