W.-T. Hsu, J. R. Clark, and C. T.-C. Nguyen, "Mechanically temperature-compensated flexural-mode micromechanical resonators," *Technical Digest*, IEEE Int. Electron Devices Meeting, San Francisco, California, Dec. 11-13, 2000, pp. 399-402.

# Mechanically Temperature-Compensated Flexural-Mode Micromechanical Resonators

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# ABSTRACT

An IC-compatible, high frequency (HF), lateral micromechanical resonator supported by a mechanical structure designed to introduce stresses that counteract temperatureinduced frequency shifts, has been demonstrated at 10 MHz with a much-reduced temperature coefficient of -2.5 ppm/°C and a Q of greater than 10,000. These values constitute substantial improvements over the -17 ppm/°C and 3,000 posted by previous clamped-clamped beam vertical resonators in this frequency range, and represent significant strides towards reducing the thermal dependence of micromechanical resonators, possibly to the point where such devices can be used in on-chip high-Q reference oscillator applications without the need for electronic temperature compensation.

### I. INTRODUCTION

Recent demonstrations of 10 MHz clamped-clamped beam vibrating micromechanical (" $\mu$ mechanical) resonators with Q's on the order of 3,000 [1] have spurred interest in using these devices as high-Q tanks for the low-phase noise reference oscillators needed in communication transceivers and instrumentation. Further incentive for doing this arises from the potential for single-chip integration of such oscillators using a merged transistor/MEMS process technology [2]. However, although the Q of such resonators should be just sufficient to attain the needed phase noise performance in this application, their thermal stability falls well short of needed specifications, typically exhibiting frequency variations on the order of 1760 ppm over a 27°C to 107°C temperature range.

In previous work [3], a geometric stress-compensation design technique was demonstrated on low-frequency (LF, e.g., 80 kHz) nickel folded-beam µmechanical resonators that used a geometrically-tailored stress versus temperature function to cancel the thermal dependence of the material Young's modulus, resulting in an overall lower frequency excursion over a given temperature range, and generating zero temperature coefficient  $TC_{fo}$  points in the process. This work now introduces a new stress-tailoring mechanical support structure to extend the above geometric stress-compensation technique to higher frequency (HF) flexural-mode beam resonators more appropriate for use in practical reference oscillator applications. Using this new design, a 10 MHz µmechanical resonator with a Q > 10,000—substantially higher than previous ones [1]—is demonstrated with a frequency variation of only 200ppm over a 27°C to 107°C temperature range.

## II. TEMPERATURE INSENSITIVE RESONATOR STRUCTURE AND OPERATION

Figure 1 presents the perspective-view schematic of the



Fig. 1: Perspective-view schematic of a mechanically temperature-compensated micromechanical resonator.

mechanically temperature-compensated resonator, indicating key components and specifying a preferred bias and excitation configuration. As shown, this structure consists of a flexuralmode resonator beam anchored to the substrate at one end, but supported on the other by a folded structure comprised of a truss section attached to two outer beams. Metal electrodes are positioned on either side of the resonator beam to allow lateral (i.e., *x*-directed) excitation using the voltage configuration shown, where a DC bias voltage  $V_P$  is applied to the conductive beam, and an ac excitation voltage  $v_i$  to the input electrode. This combination of voltages across the input electrodeto-resonator gap generates a  $V_P$ -multiplied force at the frequency of  $v_i$  that drives the beam into vibration when its frequency matches the resonance frequency  $f_o$ .

Unlike previous HF vertical resonators (i.e., moving perpendicular to the substrate), which were generally constrained by processing complexity considerations to be one-port devices, the device of Fig. 1 has two-ports. This feature, made possible by lateral operation, not only facilitates the measurement of frequency characteristics for this device by eliminating the need for a bias tee and greatly reducing port-to-port feedthrough, but also simplifies the design of systems using this device, such as oscillators [4] or *Q*-controlled filters [5].

To insure that only the resonator beam vibrates when excitation signals are applied, the outer support beams are made much wider than the resonator beam, making them rigid against lateral motions. The outer beams are also designed to be longer than the resonator beam, so they will expand faster than it with increasing temperature, generating a net tension in the resonator beam, as shown in the FEM simulation of Fig. 2.



Fig. 2: FEM-simulated Stress distribution on a temperature-insensitive resonator at an elevated temperature.

This tensile stress serves to increase the beam's resonance frequency, and thus, oppose frequency decreases caused by Young's modulus temperature dependence, resulting in a smaller overall  $f_o$  excursion over a given temperature range.

In addition to temperature insensitivity, this design also features a support structure that effectively reduces energy radiation into the substrate during resonance vibration. In particular, although one end of the resonator beam is rigidly anchored to the substrate (i.e., a fixed end condition) the other is held relatively softly by the folding truss support system, in what effectively amounts to an end condition that is somewhat more decoupled from the substrate than a directly anchored end condition. This end condition, together with the use of a lateral (rather than a vertical) vibration mode, effectively reduces anchor losses to the substrate, allowing higher resonator Q than previous clamped-clamped beam resonators.

# III. TEMPERATURE-INSENSITIVE RESONATOR DESIGN

In the lateral vibration mode, the beam of Fig. 1 essentially has fixed-fixed end conditions, so the expression for its resonance frequency is identical to that of previous clampedclamped beam resonators [1]. Using Euler-Bernoulli theory, and modeling the thermal dependence of the beam by explicitly expressing stresses due to geometric expansions or contractions as functions of temperature, the expression for resonance frequency can be written as

$$f_o = 1.027 \sqrt{\frac{E}{\rho}} \frac{W_1}{L_1^2} \left( 1 + 0.293 \frac{L_1^2}{EW_1^2} \sigma \right)^{\frac{1}{2}}$$
(1)

where  $L_1$  and  $W_1$  are the length and width of the resonator beam, respectively; and E and  $\rho$ , are the Young's modulus and density of its structural material, respectively;  $\sigma$  is the longitudinal stress generated by thermal expansion differences between the resonator beam and its compensating suspension network, given by

$$\sigma = E(\alpha_{poly} - \alpha_{sub})\Delta T \left(\frac{L_2}{L_1} - 1\right)$$
(2)

where  $\alpha_{poly}$  and  $\alpha_{sub}$  are the thermal expansion coefficients for the (polysilicon) structural material and the substrate, respectively, and  $\Delta T = T \cdot T_o$  is the temperature *T* variation from room temperature  $T_o$ . From (1) and (2), the temperature coefficient of the resonance frequency  $TC_{fo}$  is found to be

**Table I: Parameters for Temperature-Insensitive Resonators** 

Parameters	Value	Units
Resonator Beam Length, $L_1$	40	μm
Resonator Beam Width, $w_1$	2	μm
Folding Truss Length, L <sub>f</sub>	80	μm
Folding Truss Width, W <sub>f</sub>	20	μm
Structural Layer Thickness, t	2.2	μm
Compensation Beam Length, $L_2$	variable	μm
Compensation Beam Width, W <sub>2</sub>	20	μm
Electrode-Resonator Gap Spacing, d	0.1	μm
Electrode Overlap Area	44	$\mu m^2$
Nominal Resonance Frequency, fo	10	MHz
Young's Modulus of Polysilicon, E	1.5e12	dyne/cm <sup>2</sup>
Temperature Coefficient of $E$ , $\alpha_E$	-46	ppm/°C
Thermal Expansion Coeff. of Poly-Si, $\alpha_{poly}$	2.9	ppm/ <sup>o</sup> C
Thermal Expansion Coeff. of Substrate, $\alpha_{sub}$	2.0	ppm/°C
Density	2.33	g/cm <sup>3</sup>

$$TC_{fo} \approx \frac{\alpha_E + \alpha_{poly}}{2} + 0.146 \left(\frac{L_1}{W_1}\right)^2 (\alpha_{poly} - \alpha_{sub}) \left(\frac{L_2}{L_1} - 1\right)$$
 (3)

where  $\alpha_E$  is temperature coefficient of the Young's modulus, and where a dependence on the ratio of the support and resonator beam lengths,  $L_2/L_1$ , is clearly seen.

The first term of (3) is merely the  $TC_{fo}$  of a clampedclamped beam, while the second term is generated by action of the compensation structure. From (3), a reduction in the  $TC_{fo}$  is possible only if (1) the thermal expansion coefficient of the structural material is different from that of the substrate; and (2)  $L_2/L_1$  is not unity. Moreover, the  $L_1/W_1$  term in (3) can be interpreted as a gain term in the expression, that amplifies or attenuates the compensation effect when  $L_1$  and  $L_2$  are unequal. In high frequency resonators, where  $L_1/W_1$  must be chosen small, larger  $L_2/L_1$  ratios are generally required. Using the formulations above, Table I,summaries the temperatureinsensitive resonator design of this work.

## **IV. FABRICATION**

As mentioned in Section II, the key feature of this resonator design that enables improved performance is its use of lateral mode vibrations, which not only facilitates the design of the temperature compensating support structure, but also reduces anchor losses to the substrate (raising its Q), and makes available two input ports (which greatly facilitates circuit design with this device). However, for impedance matching reasons [1], lateral operation at this frequency range also necessitates high aspect-ratio, submicron, lateral electrode-toresonator gaps.

Achieving such gaps in a lateral direction with the needed aspect-ratios would be a challenging proposition for a conventional surface micromachining process armed with only present-day lithographic and etch capabilities. For this reason, this work utilizes the novel process flow summarized in Fig. 3, that combines surface-micromachining with metal electroplating technology to achieve 1000Å lateral gaps defined by a sidewall sacrificial oxide layer. In this process, a surface micromachining process is first used to achieve polysilicon structures over a bottom sacrificial oxide layer. Another oxide layer is then conformally deposited to form



Fig. 3: Process flow of the lateral thin gap process for temperature compensated micromechanical resonators.



Fig. 4: SEM of a temperature-insensitive micromechanical resonator fabricated using the process of Fig. 3.

sacrificial oxide sidewalls along the resonator beam sides (c.f., Fig. 3(a)). Gold electroplating is then performed through a mold defined by photoresist on one side, and the sidewall oxide on the other, resulting in Au electrodes separated from the resonator sides by only the sidewall oxide thickness (c.f., Fig. 3(d)). A final release step then yields the final cross-section shown in Fig. 3(e) and structures such as shown in the scanning electron micrographs (SEM's) of Fig. 4

## V. EXPERIMENTAL RESULTS

Frequency spectra and thermal dependence data were



Fig. 5: Measured spectra for (a) a lateral temperature-insensitive resonator (i.e., Fig. 4), and (b) vertical clamped-clamped beam.

obtained using the set-up shown in Fig. 1. Here, an HP 8714C Network Analyzer is used to measure transmission spectra for devices mounted on a printed circuit (PC) board contained within a custom-built vacuum chamber capable of pressures as low as 50  $\mu$ Torr. The PC board features a special thermally conductive slot to allow mounting of the resonator die on the PC board while also making sufficient thermal contact to an MMR temperature controllable cantilever.

Figure 5 compares measured frequency characteristics for two polysilicon  $\mu$ mechanical resonators: (a) a lateral temperature-insensitive resonator of this work; and (b) a vertical resonator achieved using a conventional surface- $\mu$ machining process with an identical polysilicon deposition recipe. As shown, the *Q* of the lateral resonator is almost 3X greater than that of the vertical resonator, most likely indicating better substrate isolation for the former, i.e., smaller anchor losses.

Figure 5 presents a plot of fractional frequency change versus temperature for resonators such as in Fig. 4 with varying  $L_2/L_1$  beam ratios (see Fig. 1). From this plot, control of the resonance frequency temperature dependence can clearly be seen. In particular, for  $L_2/L_1$  beam ratios from 30/40 to 60/ 40, the temperature coefficient decreases from -35.9 ppm/°C for the  $L_2/L_1=30\mu m/40\mu m$  device to -2.5 ppm/°C for the  $L_2/L_1=60\mu m/40\mu m$  device. Unfortunately, however, for higher beam ratios, the temperature coefficient is seen to degrade. A possible mechanism for this is one where the support beams bend upwards or downwards when the stress in the beams becomes too large in devices with large  $L_2/L_1$  beam ratios,



Fig. 6: Fractional frequency change versus temperature for the device of Fig. 4 with various compensation ratios.



Fig. 7: FEM vertical displacement simulations for (a) conventional anchors and (b) side-anchors. Beam bending is seen in (a).

thereby suppressing and eventually reversing tensile stresses.

That bending occurs in the structure has now been verified via FEM analysis. Figure 7(a) presents the FEM-simulated shape of a temperature-insensitive resonator with the design of Fig. 1 at an elevated temperature of 100°C, clearly showing an upward bending of the beam. However, simulation also shows that this bending occurs even for a cantilever beam with a rigid anchor of the type used in Fig. 7(a). This is immediately understandable upon recognition that in this anchor design, the bottom side of the anchor is attached to the substrate, so it expands or contracts at a different rate than its top side, and this results in downward bending of the beam, as shown in Fig. 7(a). Although, on one hand, this result seems to lend credence to a bending-based mechanism for the observed back-and-forth  $TC_{fo}$  behavior of Fig. 6, on the other hand, it raises questions as to whether or not the mechanism behind the initial lowering of the  $TC_{fo}$  is really as described in Section II. In other words, does  $TC_{fo}$  control occur due to thermal expansion differences that induce stress directly in the beam, or does control actually arise less directly, from expansion differences that induce stresses to cause beam bending.

Pursuant to answering this question, a second fabrication run was done that included structures using a modified anchor design that eliminates the beam bending mechanism described above. This new anchor, shown in Fig. 7(b), redirects top-tobottom side expansion differences so that they occur only in a direction orthogonal to the length of the resonator (or support) beam, thus, eliminating beam bending along the length of the



beam. Figure 8 presents the SEM of a 10 MHz version of a temperature-insensitive resonator using these new anchors. The inset of Fig. 8 presents the fractional frequency versus temperature curve for this resonator, again for various values of  $L_2/L_1$ , showing clearly that without bending, there is very little control over the frequency versus temperature curve. This suggests that the difference in the thermal expansion coefficients of substrate silicon and polysilicon is large enough to cause beam bending, but not large enough to generate enough direct beam stress to alter its  $TC_{for}$ .

Next generation structures are presently underway that use different materials for the resonator and stress-controlling suspension beams to insure the *direct* generation of sufficient stresses.

#### VI. CONCLUSIONS

Using a new geometric stress-compensating support structure together with a process technology that achieves submicron lateral electrode-to-resonator gaps, micromechanical resonators at HF frequencies have been demonstrated with both higher Q's and lower temperature coefficients than previous clamped-clamped beam resonators in this frequency range. Although in this work the mechanism for temperature compensation may not have been as direct as originally expected by design (due to beam bending effects), the concept of using a geometrically-tailored stress versus temperature function to compensate for Young's modulus thermal dependence still applied, and should prove instrumental in future resonator designs targeted for communications and timekeeper applications.

*Acknowledgment*: This work was supported under DARPA Cooperative Agmt. No. F30602-97-2-0101.

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